

# On the Sample Complexity of Stabilizing LTI Systems on a Single Trajectory

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## Problem Setting

We consider a noiseless LTI system

$$x_{t+1} = Ax_t + Bu_t,$$

where:

- $x_t \in \mathbb{R}^n$  is the **state**, and  $u_t \in \mathbb{R}^m$  is the **control input**.
- **Dynamical matrices**  $A \in \mathbb{R}^{n \times n}$  and  $B \in \mathbb{R}^{n \times m}$  are unknown.
- The learner is allowed to learn about the system by interacting with it on a **single trajectory**.
  - The initial state is sampled from the unit hyper-sphere surface in  $\mathbb{R}^n$  uniformly at random.
  - At each time step  $t$ , the learner is allowed to observe  $x_t$  and freely determine  $u_t$ .
- The goal of the learner is to learn a **stabilizing controller**.

We make the following assumptions on the system.

**Assumption 1** (Spectral Property).  $A$  is diagonalizable with **instability index**  $k$  and **distinct eigenvalues**  $\lambda_1, \dots, \lambda_n$  satisfying  $|\lambda_1| \geq |\lambda_2| \geq \dots \geq |\lambda_k| > 1 > |\lambda_{k+1}| \geq \dots \geq |\lambda_n|$ .

**Assumption 2** (Initialization). The initial state of the system is sampled uniformly at random on the unit hyper-sphere surface in  $\mathbb{R}^n$ .

**Assumption 3** ( $(\nu, \sigma)$ -Strong Controllability). The system is  $(\nu, \sigma)$ -strongly controllable; i.e.,  $\sigma_{\min}(C_\nu) > \sigma$ , where  $C_\nu := [A^{\nu-1}B \ A^{\nu-2}B \ \dots \ AB \ B]$  is the  $\nu$ -step controllability matrix.

## Subspace Decomposition

Under Assumption 1, define the subspaces:

- the **eigenspace**  $E_i = \text{span}(v_i)$  for an eigenvector  $v_i$  corresponding to  $\lambda_i$ , define
- the **unstable subspace**  $E_u := \bigoplus_{i \leq k} E_i$ ;
- the **stable subspace**  $E_s := \bigoplus_{i > k} E_i$ .

Suppose the unstable subspace  $E_u$  and its orthogonal complement  $E_u^\perp$  are spanned by **orthonormal** columns of  $P_1 \in \mathbb{R}^{n \times k}$  and  $P_2 \in \mathbb{R}^{n \times (n-k)}$ , respectively, namely

$$E_u = \text{col}(P_1), \quad E_u^\perp = \text{col}(P_2).$$

Write  $P = [P_1 \ P_2]$  as a shorthand. Let  $\Pi_1 := P_1 P_1^\top$  and  $\Pi_2 = P_2 P_2^\top$  be the **orthogonal projectors** onto  $E_u$  and  $E_u^\perp$ , respectively.

Under such subspace decomposition, we shall also decompose  $A$  as

$$AP = P \begin{bmatrix} M_1 & \Delta \\ & M_2 \end{bmatrix} \Leftrightarrow M := \begin{bmatrix} M_1 & \Delta \\ & M_2 \end{bmatrix} = P^{-1}AP.$$

- The top-left block  $M_1$  represents the action of  $A$  on the unstable subspace.
- **This is the "small part" that leads to instability which we want to eliminate.**
- The matrices  $M_1$  (and  $P_1$ ), as compared to  $A$ , is much smaller in size and thus **takes much fewer samples to learn**.

**How to learn about the unstable subspace?** A critical observation is that, when we recursively apply  $A$  to a state vector, the stable component of it will shrink, and the unstable component will stretch. Therefore, after letting the system run in open loop for sufficiently many heat-up steps, **the state vector is automatically pushed to approach  $E_u$**  (see the top right figure).

We propose a novel algorithm to learn to stabilize an unknown linear time-invariant (LTI) system on a single trajectory. It uses  **$O(k \log n)$  samples** to stabilize a system with instability index  $k$ , which is **sublinear** in  $n$  when  $k \ll n$ .

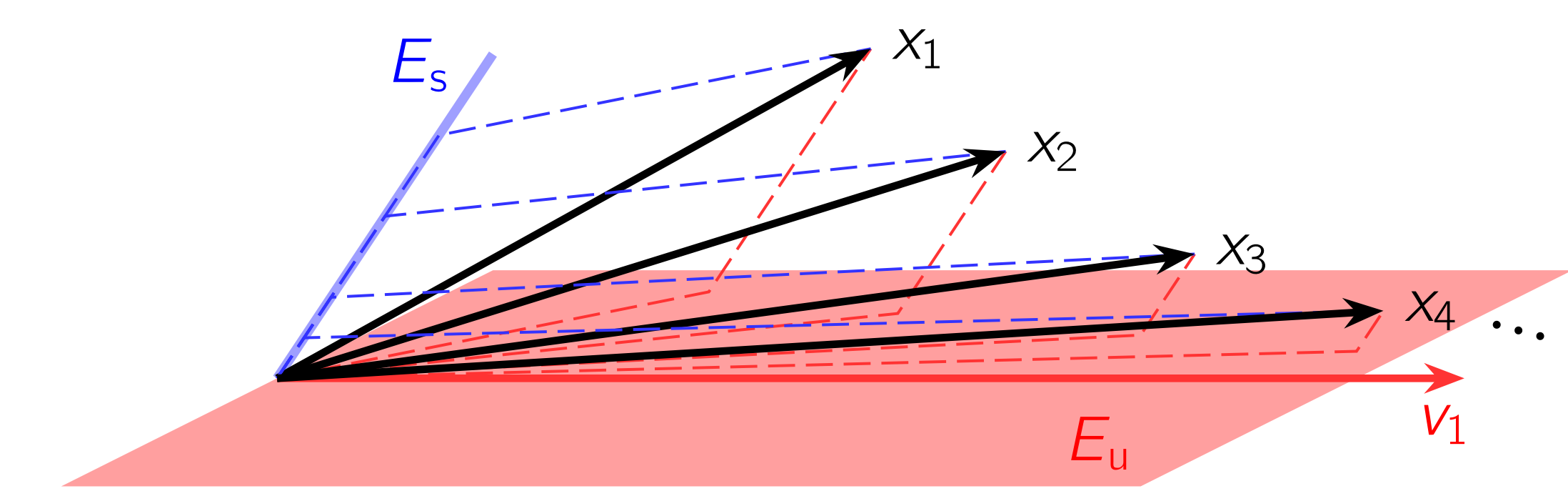
Learn  $E_u = \text{col}(P_1)$ .

Learn dynamics  $(A, B)$  "restricted" to  $E_u$ .

Stabilize the "restricted" system every  $\tau$  steps.



Scan the QR code to download the full paper



System state evolution in open loop.

## Algorithm Design

### Algorithm 1 LTS<sub>0</sub>: Learning a $\tau$ -hop Stabilizing Controller

**Stage 1: learn the unstable subspace of  $A$ .**

Run the system in open loop for  $t_0$  steps for initialization.

Run the system in open loop for  $k$  more steps and let  $D \leftarrow [x_{t_0+1} \ \dots \ x_{t_0+k}]$ .

Calculate  $\hat{\Pi}_1 \leftarrow D(D^\top D)^{-1}D^\top$ .

Calculate the top  $k$  (normalized) eigenvectors  $\hat{v}_1, \dots, \hat{v}_k$  of  $\hat{\Pi}_1$ , and let  $\hat{P}_1 \leftarrow [\hat{v}_1 \ \dots \ \hat{v}_k]$ .

**Stage 2: approximate  $M_1$  on the unstable subspace.**

Solve the least squares  $\hat{M}_1 \leftarrow \arg \min_{M_1 \in \mathbb{R}^{k \times k}} \mathcal{L}(M_1) := \sum_{t=t_0+1}^{t_0+k} \|\hat{P}_1^\top x_{t+1} - \hat{M}_1 \hat{P}_1^\top x_t\|^2$ .

**Stage 3: restore  $B_\tau$  for  $\tau$ -hop control.**

for  $i \leftarrow 1, \dots, k$  do

Let the system run in open loop for  $\omega$  time steps.

Run for  $\tau$  more steps with initial  $u_{t_i} = \alpha \|x_{t_i}\| e_i$ , where  $t_i = t_0 + k + i\omega + (i-1)\tau$ .

Let  $\hat{B}_\tau \leftarrow [\hat{b}_1 \ \dots \ \hat{b}_k]$ , where the  $i^{\text{th}}$  column  $\hat{b}_i \leftarrow \frac{1}{\alpha \|x_{t_i}\|} (\hat{P}_1^\top x_{t_i+\tau} - \hat{M}_1^\tau \hat{P}_1^\top x_{t_i})$ .

**Stage 4: construct a  $\tau$ -hop stabilizing controller  $K$ .**

Construct the  $\tau$ -hop stabilizing controller  $\hat{K} \leftarrow -\hat{B}_\tau^{-1} \hat{M}_1^\tau \hat{P}_1^\top$ .

**$\tau$ -hop control.** The intuition of stabilizing the unstable component only holds when the state is close enough to the unstable subspace. Luckily, the stable component automatically vanishes over time, so we design the controller to eliminate the unstable component **only "once in a while"**. The closed-loop dynamics with the  $\tau$ -hop controller can be written as

$$\tilde{y}_{s+1} = \begin{bmatrix} M_1^\tau + P_1^\top A^{\tau-1} B \hat{K}_1 \hat{P}_1^\top P_1 & \Delta_\tau + P_1^\top A^{\tau-1} B \hat{K}_1 \hat{P}_1^\top P_2 \\ P_2^\top A^{\tau-1} B \hat{K}_1 \hat{P}_1^\top P_1 & M_2^\tau + P_2^\top A^{\tau-1} B \hat{K}_1 \hat{P}_1^\top P_2 \end{bmatrix} \begin{bmatrix} \tilde{y}_{1,s} \\ \tilde{y}_{2,s} \end{bmatrix}.$$

## Performance of the Algorithm

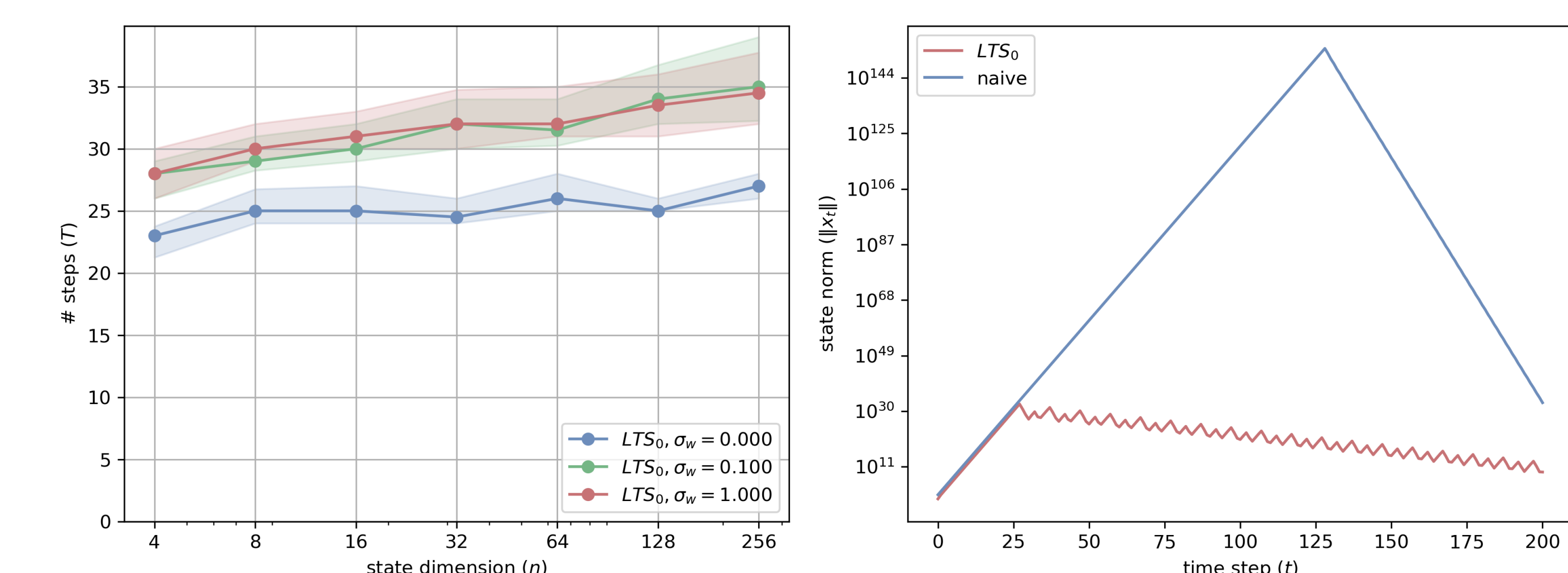
**Theoretical guarantee.** We show the following Main Theorem to provide theoretical performance guarantee for LTS<sub>0</sub> in noiseless LTI systems.

**Theorem 1** (Main Theorem). Given a noiseless LTI system  $x_{t+1} = Ax_t + Bu_t$  subject to Assumptions 1, 2 and 3, and additionally  $|\lambda_1|^2 |\lambda_{k+1}| < |\lambda_k|$ , by running LTS<sub>0</sub> with parameters

$$\tau = O(1), \quad \omega = O(\ell \log k), \quad \alpha = O(1), \quad t_0 = O(k \log n)$$

that terminates within  $t_0 + k(1 + \omega + \tau) = O(k \log n)$  time steps, the closed-loop system is exponentially stable with probability  $1 - O(k^{-\ell})$  over the initialization of  $x_0$  for any  $\ell \in \mathbb{N}$ . Here the big- $O$  notation only shows dependence on  $k$  and  $n$  and hides dependency on instance-specific parameters.

**Experimental results.** Though for clarity of exposition our Main Theorem does not contain disturbances, we show by numerical experiments that our algorithm LTS<sub>0</sub> can also handle disturbances quite well.



(a) Running steps of LTS<sub>0</sub>

(b) State norms along one trajectory